



广东工业大学
Guangdong University of Technology

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通信电路与系统

信息工程学院

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第五章 正弦波振荡器

信息工程学院

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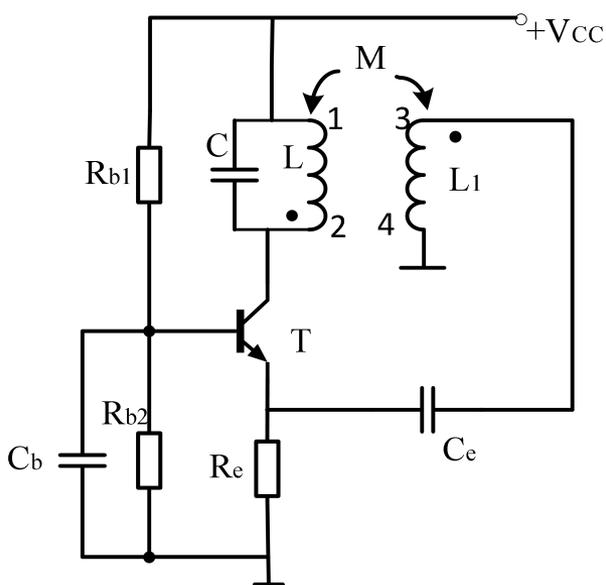
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5.2 反馈型LC振荡电路

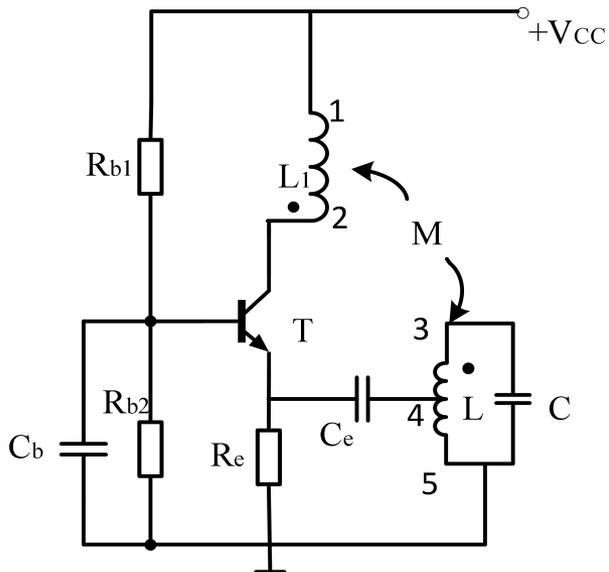
一、互感耦合振荡电路

1. 反馈振荡器的组成



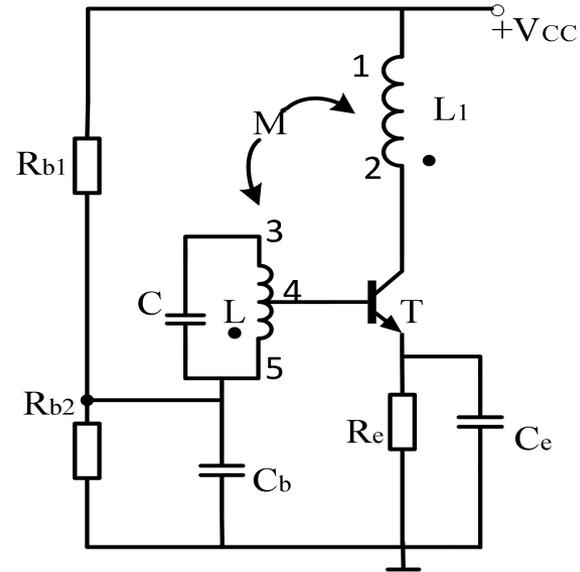
(a) 集电极调谐

调集



(b) 发射极极调谐

调射



(c) 基极调谐

调基

互感耦合式振荡器是利用电感之间的耦合来实现正反馈的。
同名端的位置决定了能否实现正反馈。

5.2 反馈型LC振荡电路

2. 调集振荡器

① 分析电路形式

一找组成，二查静态、动态

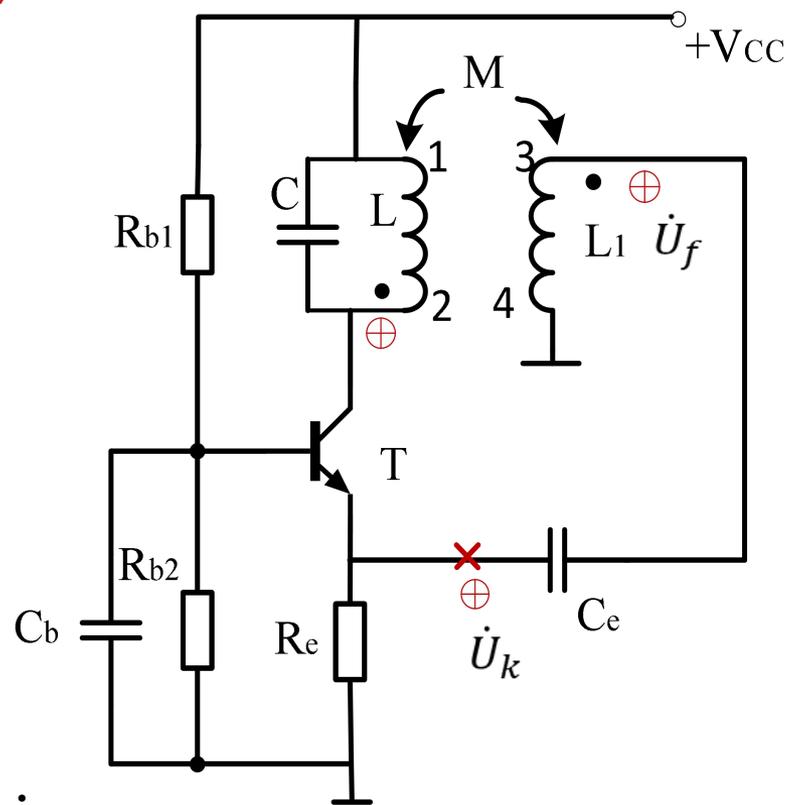
② 判断振荡条件

相位条件：采用瞬时极性法。

$\dot{U}_k \oplus \rightarrow \dot{U}_c \oplus \rightarrow$ 由同名端可知3端为 $\oplus \rightarrow \dot{U}_f \oplus$

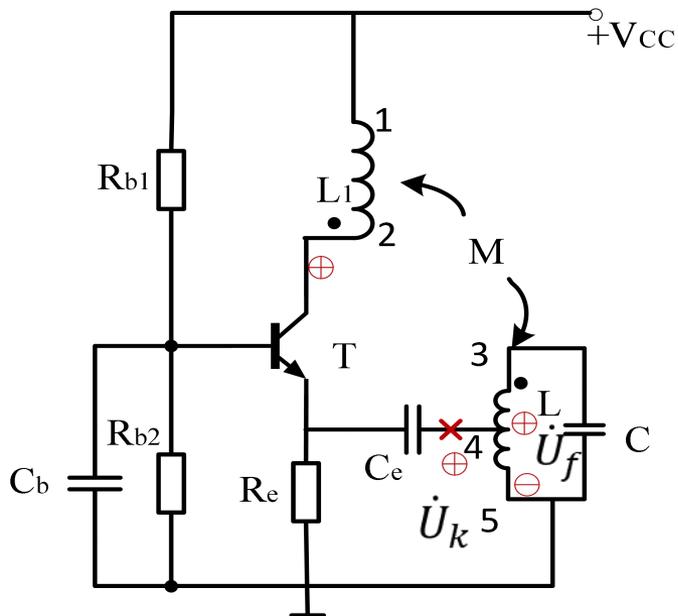
③ 估算振荡频率

由选频回路决定。 $f_0 = \frac{1}{2\pi\sqrt{LC}}$



5.2 反馈型LC振荡电路

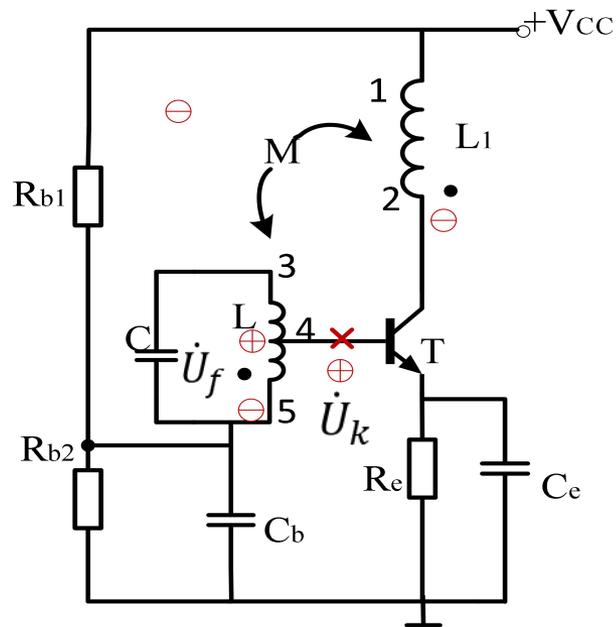
3. 调射振荡器



发射极调谐

\dot{U}_f 取自 L_{45} , 加在发射极, 共基组态

$$\dot{U}_k \oplus \rightarrow \dot{U}_c \oplus \rightarrow \text{4端} \oplus \rightarrow \dot{U}_f \oplus$$



基极调谐

\dot{U}_f 取自 L_{45} , 加在基极, 共射组态

$$\dot{U}_k \oplus \rightarrow \dot{U}_c \oplus \rightarrow \text{4端} \oplus \rightarrow \dot{U}_f \oplus$$

注意：反馈元件的一个端交流接地

5.2 反馈型LC振荡电路

二、三端式振荡器的相位平衡条件

$$\varphi_{\Sigma} = 2n\pi \quad n = 0, 1, 2, \dots$$

$$F = \frac{U_f}{U_0} = \frac{i_x X_{be}}{-i_x X_{ce}} = -\frac{X_{be}}{X_{ce}}$$

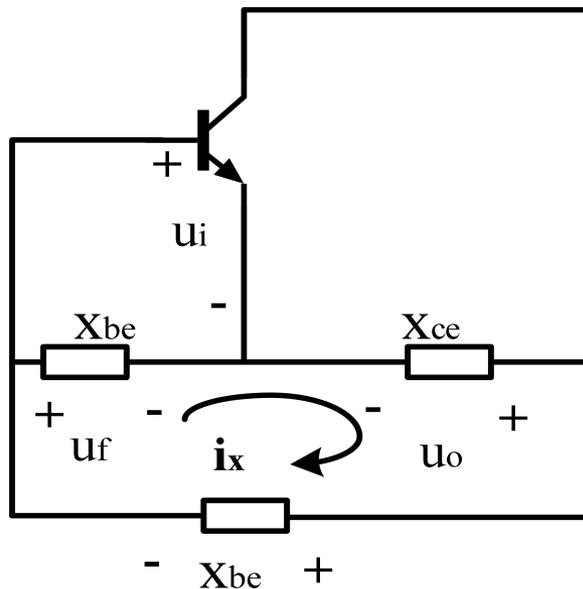
正反馈 $\dot{U}_i \xrightarrow{\ominus} \dot{U}_0 \xrightarrow{\ominus} \dot{U}_f$

$$\therefore \frac{X_{be}}{X_{ce}} \geq 0$$

X_{be} 、 X_{ce} 为同性电抗元件

$$\omega = \omega_0 \quad \Sigma X = X_{be} + X_{ce} + X_{bc} = 0$$

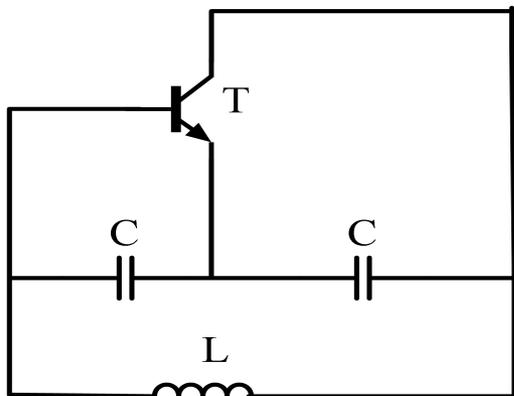
$$X_{bc} = -(X_{be} + X_{ce}) \quad X_{be} \text{ 与 } X_{ce}、X_{bc} \text{ 异性}$$



X_{be} 、 X_{ce} 、 X_{bc} 均为电抗元件

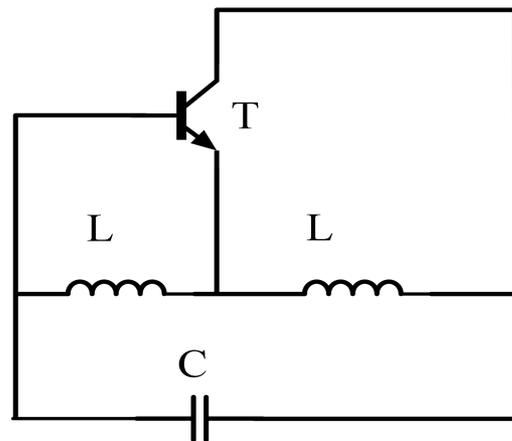
射同基反

5.2 反馈型LC振荡电路



$$X_{bc} : L; X_{be}, X_{ce} : C$$

电容反馈三点式



$$X_{bc} : C; X_{be}, X_{ce} : L$$

电感反馈三点式

源同栅反 同相端相同，反向端相反

5.2 反馈型LC振荡电路

例

B反: L, L_1 并 C_1 为 C

$$\omega > \omega_{01} = \frac{1}{\sqrt{L_1 C_1}}$$

E同: L_1 并 C_1 为 C

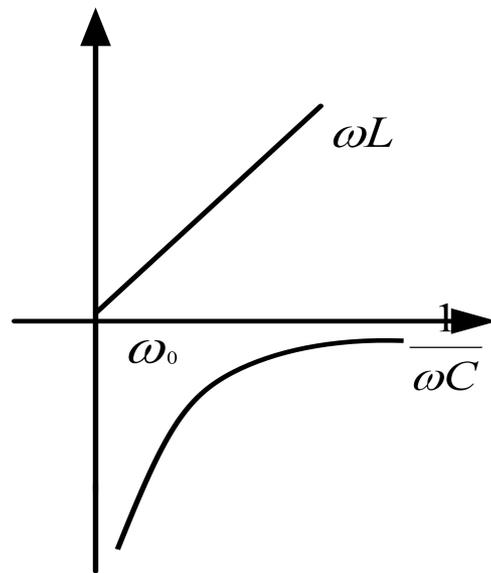
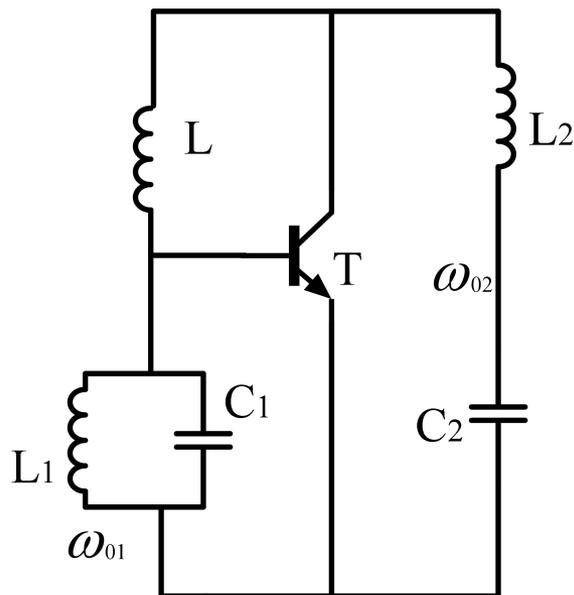
L_2 并 C_2 为 C

$$\omega < \omega_{02} = \frac{1}{\sqrt{L_2 C_2}}$$

$$\therefore \omega_{01} < \omega < \omega_{02}$$

$$\frac{1}{\sqrt{L_1 C_1}} < \omega < \frac{1}{\sqrt{L_2 C_2}}$$

$$L_1 C_1 > L_2 C_2$$



5.2 反馈型LC振荡电路

三、三端式振荡器主要参数的计算

1. 谐振频率计算

$$\omega = \omega_0 \quad \text{谐振}$$

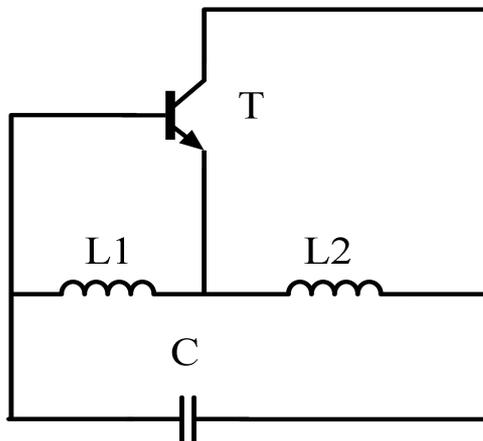
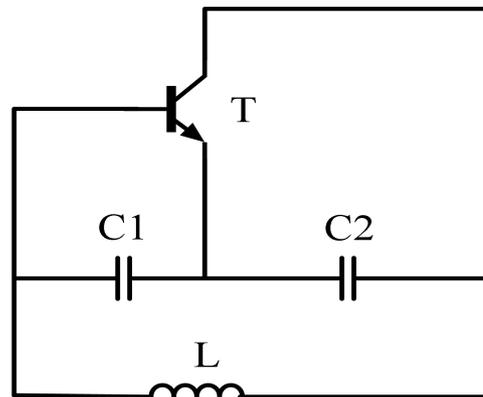
$$\frac{1}{j\omega_0 C_1} + \frac{1}{j\omega_0 C_2} + j\omega_0 L = 0$$

$$j\left(-\frac{1}{\omega_0 C_1} - \frac{1}{\omega_0 C_2} + \omega_0 L\right) = 0$$

$$\omega_0 L = \frac{1}{\omega_0 C_1} + \frac{1}{\omega_0 C_2} \quad \omega_0^2 = \frac{1}{L} \left(\frac{1}{C_1} + \frac{1}{C_2}\right) = \frac{1}{LC} \quad \frac{1}{C} = \frac{1}{C_1} + \frac{1}{C_2}$$

$$\omega_g = \omega_0 = \frac{1}{\sqrt{LC_\Sigma}} \quad C_\Sigma = C_1 // C_2$$

$$\omega_g = \omega_0 = \frac{1}{\sqrt{LC_\Sigma}} = \frac{1}{\sqrt{C(L_1 + L_2)}}$$



5.2 反馈型LC振荡电路

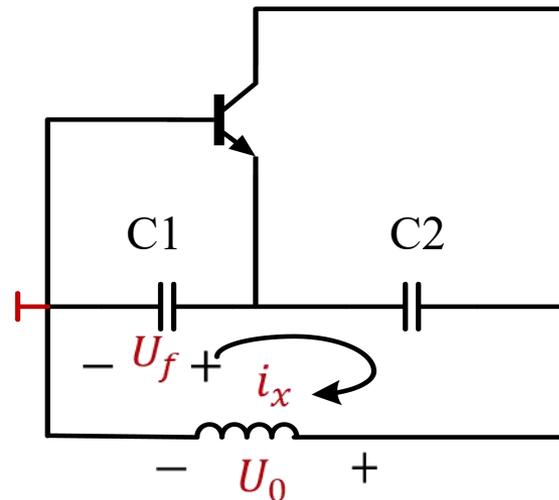
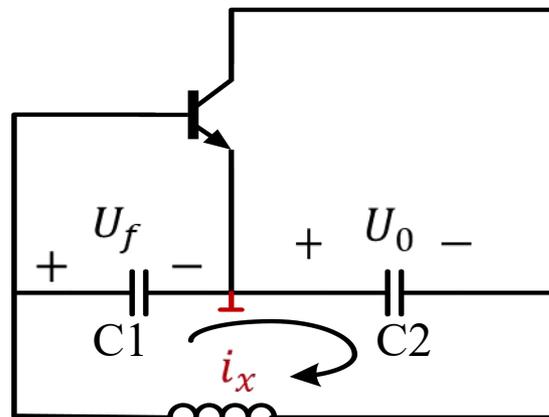
2. 反馈系数的计算

(1) 电容三端式

CE:
$$F = \frac{U_f}{U_0} = \frac{C_2}{C_1}$$

CB:
$$F = \frac{U_f}{U_0} = \frac{i_x \frac{1}{j\omega_0 C_1}}{i_x j\omega_0 L} = \frac{C}{C_1}$$

$$= \frac{C_1 C_2}{C_1 (C_1 + C_2)} = \frac{C_2}{C_1 + C_2}$$



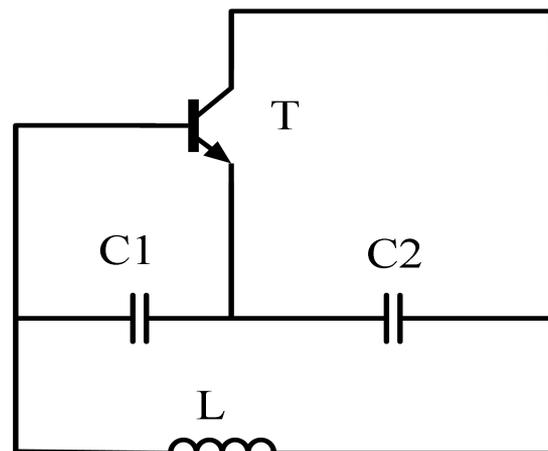
5.2 反馈型LC振荡电路

$$\omega_g = \omega_0 = \frac{1}{\sqrt{L(C_1 // C_2)}}$$

$$\text{CE: } F = \frac{C_2}{C_1} \quad \text{CB: } F = \frac{C_2}{C_1 + C_2}$$

高频时考虑 $C_{b'c} \downarrow C_{b'e} C_{ce}$

$$C_1' = C_1 + C_{b'e} \quad C_2' = C_2 + C_{ce}$$



优点： 适合高频，可工作于Microwave
频率稳定度高；波形较好

缺点： 起振较难

5.2 反馈型LC振荡电路

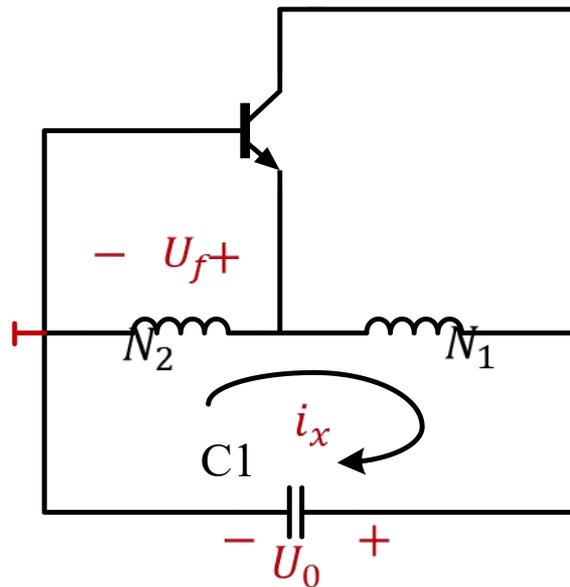
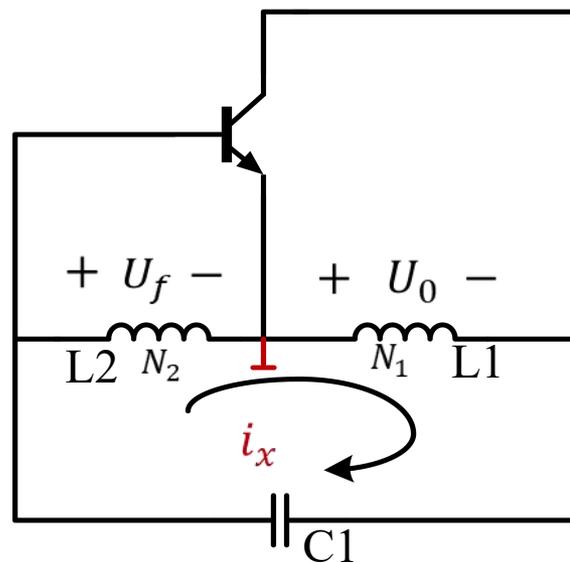
(2) 电感三点式

$$\omega_g = \omega_0 = \frac{1}{\sqrt{C(L_1 + L_2)}}$$

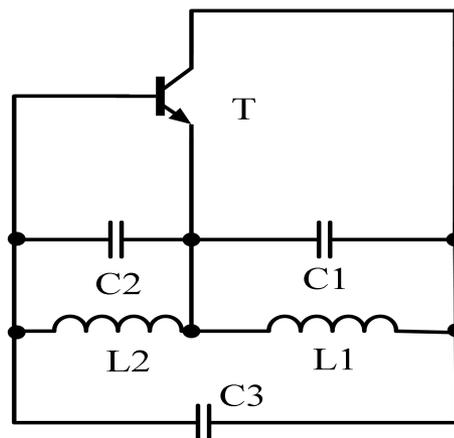
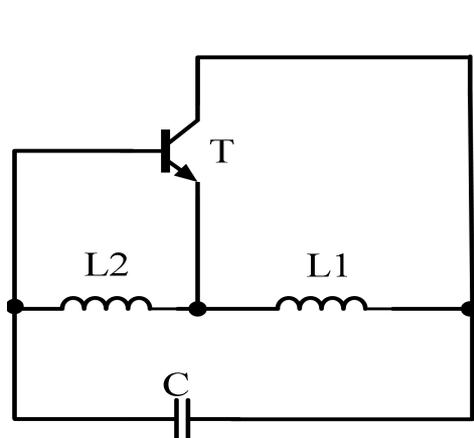
CE: $F_{ce} = \frac{U_f}{U_0} = \frac{L_2}{L_1} = \frac{N_2}{N_1}$

考虑互感 $F_{ce} = \frac{L_2 + M}{L_1 + M}$

CB: $F_{cb} = \frac{U_f}{U_0} = \frac{L_2}{L_1 + L_2} = \frac{N_2}{N_1 + N_2}$



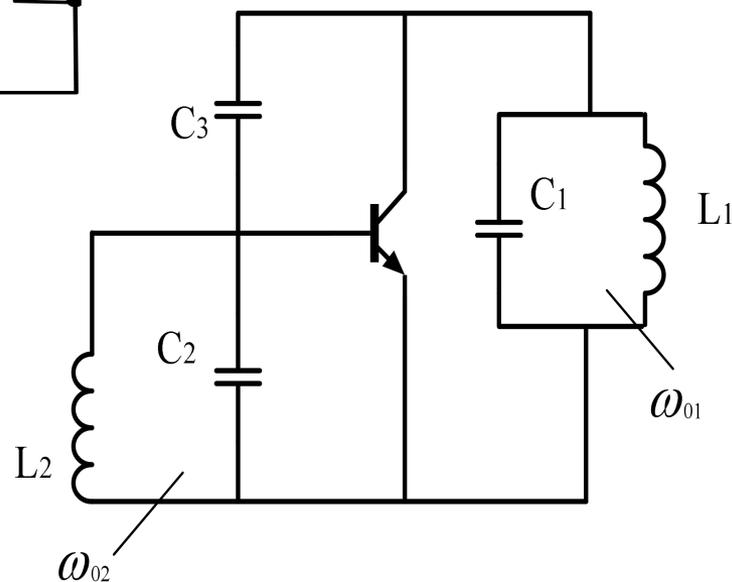
5.2 反馈型LC振荡电路



高频时考虑 $C_{b'c} \downarrow C_{b'e} C_{ce}$

$$L_1' = L_1 // C_{ce}$$

$$L_2' = L_2 // C_{b'e}$$



优点： 易起振

$$\omega < \min(\omega_{01}, \omega_{02})$$

缺点： 不适于高频；频率稳定度较差；波形较差