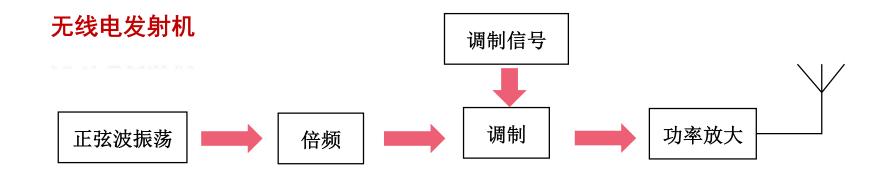
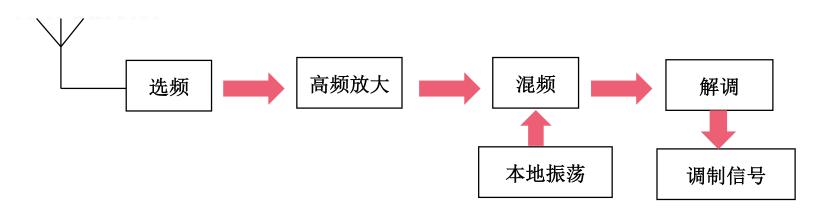


目录

- 5.1 概述与反馈型 LC 振荡原理
- 5.2 反馈型 LC 振荡电路
- 5.3 振荡器的频率稳定原理和高稳定度的 LC 振荡器
- 5.4 晶体振荡电路



无线电接收机



振荡器?

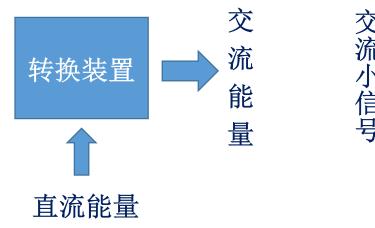
振荡器(oscillator):没有输入信号,只有直流电源供电,就可以产生并输出一定波形、一定频率和一定功率的交流信号的电路。

振荡器也可以看作是将直流电源能量转换为交流电能量的装置



振荡器与放大器的区别

振荡器无信号输入 自己产生交流信号 放大器有信号输入 将输入的交流信号放大

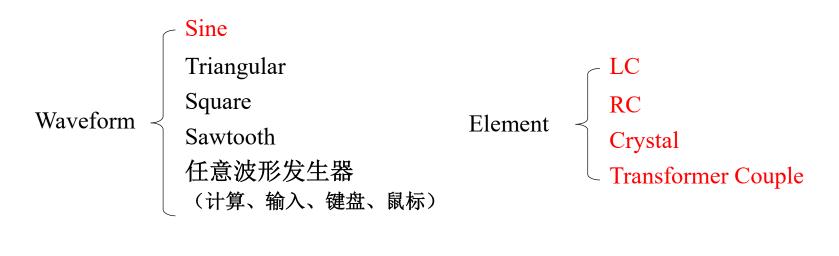


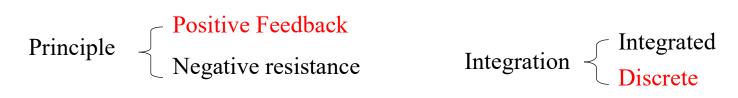
交流小信号 放大器 立流 大信号 直流能量

振荡器无中生有

放大器照猫画虎

—. Classification



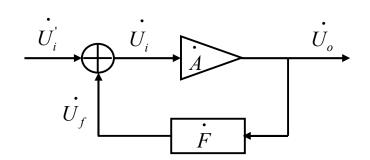


Use Power: MicroWave oven、医疗电子设备
Frequency: Watch、Carrier Wave、Local Oscillator

◆ 5.1 反馈式振荡器的工作原理

$$\dot{U}_o = \dot{A}\dot{U}_i$$
 $\dot{U}_f = \dot{F}\dot{U}_o$

$$\dot{U}_o = \dot{A} \left(\dot{U}_i' - \dot{U}_f \right) = \dot{A} \dot{U}_i' - \dot{A} \dot{F} \dot{U}_o$$



$$\therefore \frac{\dot{U}_o}{\dot{V}_i} = \frac{\dot{A}}{1 + AF}$$
 Negative Feedback Amplifier

$$\begin{array}{cccc}
& \dot{U}_o = \dot{A}\dot{U}_i' + \dot{A}\dot{F}\dot{U}_o \\
& \dot{U}_i' = \frac{\dot{A}}{1 - \dot{A}\dot{F}}
\end{array}$$

$$\dot{U}_{o} = \frac{\dot{A}}{\dot{U}_{i}} = \frac{\dot{A}}{1 - \dot{A}F}$$

$$\dot{A}\dot{F} = 1 \qquad \frac{\dot{U}_o}{\dot{U}_i} = \infty$$

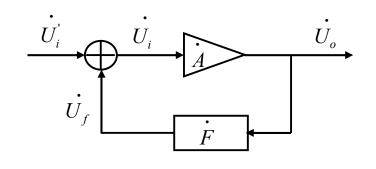
$$\dot{U}_i'$$

◆ 5.1.1 平衡条件

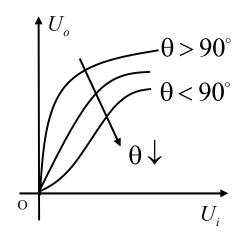
$$\dot{A}\dot{F} = 1 = \frac{\dot{U_o}}{\dot{\cdot}} \cdot \frac{\dot{U_f}}{\dot{\cdot}} = \frac{\dot{U_f}}{\dot{\cdot}}$$

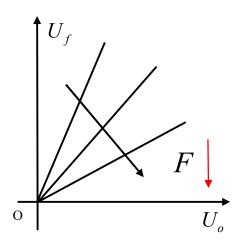
$$\dot{U_i} \quad \dot{U_o} \quad \dot{U_i}$$

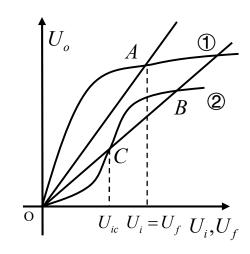
$$\begin{cases} \dot{U_f} = \dot{U_i} \\ \vdots \\ \dot{U_i} = \dot{U_i'} + \dot{U_f} \end{cases} \qquad \dot{U_i'} = 0$$

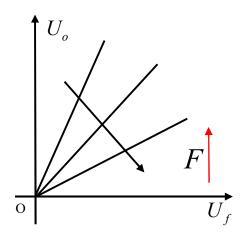


◆ 1. 振幅平衡条件



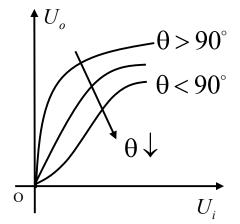


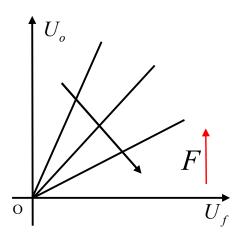


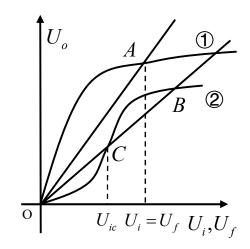


$$AF = 1 \Leftrightarrow U_f = U_i$$
 振幅平衡条件

振幅平衡条件: 反馈电压等于输入电压幅值; $U_f = U_i$ 的点即为满足振幅平衡条件的平衡点, 相应的 U_o ,就是振荡器产生的电压振幅







$$\begin{vmatrix}
\dot{U}_{o} = \dot{A}\dot{U}_{i} \\
\dot{U}_{o} = \frac{\dot{U}_{f}}{\dot{F}}
\end{vmatrix}$$

$$\begin{vmatrix}
\dot{A}\dot{F}\dot{U}_{i} = \dot{U}_{f} \\
\dot{U}_{i} = \dot{U}_{f}
\end{vmatrix}$$

$$AF = 1$$

◆ 2. 相位平衡条件

$$\varphi_A + \varphi_F = 2n\pi$$
 反馈电压 U_f 与输入电压 U_i 同相,即正反馈

$$\dot{U}_o = \dot{I}_L \dot{Z}_L(\omega) = \dot{g}_m \dot{U}_i \dot{Z}_L(\omega)$$

$$A = \frac{\dot{U}_o}{\dot{U}_i} = \frac{\dot{g}_m \dot{U}_i \dot{Z}_L(\omega)}{U_i} = \dot{g}_m \dot{Z}_L(\omega) = Ae^{i\varphi_A}$$

$$f_0 \ll f_T \qquad g_m \ll g_m \qquad \varphi_Y = 0$$

$$n = 0$$
 $\varphi_A + \varphi_F = 0$ $\therefore \varphi_A = -\varphi_F$

$$\varphi_A + \varphi_F = \varphi_Y + \varphi_Z + \varphi_F = 0$$

若令
$$\varphi_Y + \varphi_F = \varphi_{YF}$$
,则 $\varphi_Z = -\varphi_{YF}$

$$\therefore \varphi_A = \varphi_Y + \varphi_Z$$

φν:集电极电流基波分量

φz:负载的相角

$$\varphi_A + \varphi_F = \varphi_Y + \varphi_Z + \varphi_F = 0 \qquad \qquad \varphi_Y + \varphi_F = \varphi_{YF} \qquad \qquad \varphi_Z = -\varphi_{YF}$$

$$\varphi_Y + \varphi_F = \varphi_{YF}$$

$$\varphi_Z = -\varphi_{YF}$$

 φ_F :

$$U_{o} = \frac{1}{U_{o}} U_{f} \qquad V_{o} = \frac{i \frac{1}{j \omega c_{2}}}{i \frac{1}{j \omega c_{2} / / c_{1}}} = \frac{c_{2} / / c_{1}}{c_{2}} = \frac{c_{1}}{c_{2} + c_{1}}$$

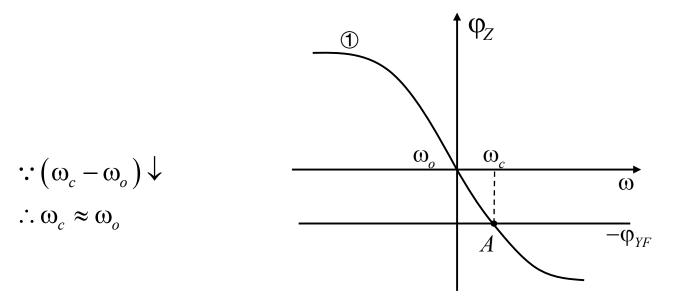
$$\therefore \varphi_F = 0$$

一般情况: 晶体管少数载流子在通过基区有效宽度时, 总需要一定的扩散时间,即 I_c 总滞后于 U_i

$$\therefore \varphi_{Y} < 0 \qquad \therefore \varphi_{YF} = \varphi_{Y} + \varphi_{F} \neq 0 \qquad \therefore \varphi_{Z} = -\varphi_{YF} \neq 0$$

:: 谐振时呈现纯阻 :: 失谐工作即: $\omega_c \neq \omega_o$

 ω_c : 电路参数决定的振荡频率 ω_c : 回路谐振频率



LC并联振荡回路 负载相角与频率的关系

◆ 5.1.2 稳定条件 stabilization Condition

1. 振幅稳定条件

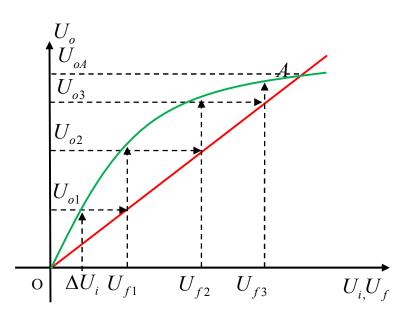
$\theta \ge 90^{\circ}$

放大特性与反馈特性有两个交点O、A

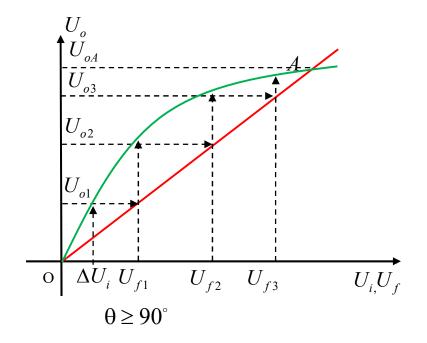
电源接通瞬间 $\dot{U}_i = 0$, $\dot{U}_o = 0$ 外界电磁感应在放大器输入端感应电压 ΔU_i ,放大器输出 U_{o1} ,经过反馈网络,反馈电压 U_{f1} ,由于 $U_{f1} > \Delta U_i$,振荡器就会脱离开原点而振荡起来。

若因外界因素振荡偏离A点,仍可返回





θ≥90°的放大特性与反馈特性

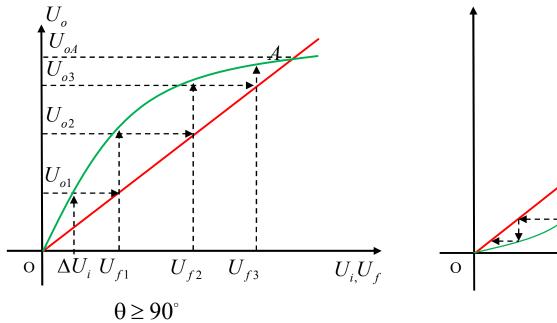


 $\theta < 90^{\circ}$

A 稳定点
Soft Self Excitation **软自激**

O、B 稳定点; C 非稳定点 初始为 O,不起振 $U_i > U_c$ 自动起振

Hard Self Excitation 硬自激



 $\theta < 90^{\circ}$

A 稳定点

O、B稳定点; C非稳定点

稳定: 放大特性斜率小于反馈特性斜率

$$\frac{\partial U_o}{\partial U_i} \mid p < \frac{\partial U_o}{\partial U_f} \mid p$$

$$\frac{\partial U_o}{\partial U_i} \frac{\partial U_f}{\partial U_o} < 1$$

$$\frac{\partial U_f}{\partial U_i} \mid p < 1$$

$$\frac{\partial U_f}{\partial U_i} \mid p < 1 \qquad \qquad \frac{\partial U_f}{\partial U_i} = \dot{F} \dot{U_i} \frac{\partial \dot{A}}{\partial \dot{U_i}} + \dot{A} \dot{U_i} \frac{\partial \dot{F}}{\partial \dot{U_i}} + \dot{A} \dot{F} \frac{\partial \dot{U}_i}{\partial \dot{U_i}} < 1$$

$$U_f = \stackrel{\cdot}{A} \stackrel{\cdot}{F} \stackrel{\cdot}{U}_i$$

平衡点:

$$\dot{A}\dot{F} = 1 \qquad \dot{F}\dot{U_i}\frac{\partial \dot{A}}{\partial \dot{U_i}} + \dot{A}\dot{U_i}\frac{\partial \dot{F}}{\partial \dot{U_i}} < 0 \qquad \dot{F}\frac{\partial \dot{A}}{\partial \dot{U_i}} + \dot{A}\frac{\partial \dot{F}}{\partial \dot{U_i}} < 0$$

$$\begin{cases} F = 常数 & \frac{\partial \dot{F}}{\partial \dot{U}_i} = 0 & \therefore \dot{F} \frac{\partial \dot{A}}{\partial \dot{U}_i} < 0 & \text{ 内稳幅条件} \\ A = 常数 & \frac{\partial \dot{A}}{\partial \dot{U}_i} = 0 & \therefore \dot{A} \frac{\partial \dot{F}}{\partial \dot{U}_i} < 0 & \text{ 外稳幅条件} \\ \frac{\partial \dot{A}}{\partial \dot{U}_i} = 0 & \therefore \dot{A} \frac{\partial \dot{F}}{\partial \dot{U}_i} < 0 & \text{ 非线性反馈网络} \end{cases}$$

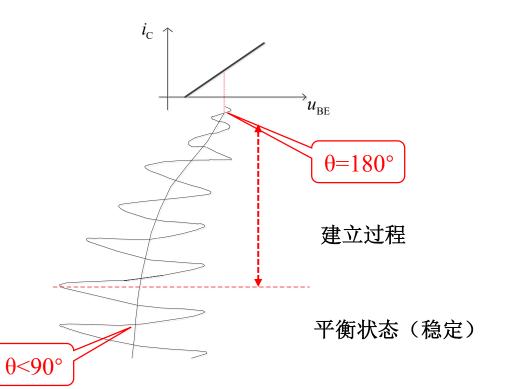
$$A = 常数$$
 $\frac{\partial \dot{A}}{\partial \dot{U}} = 0$ $\therefore \dot{A} \frac{\partial \dot{F}}{\partial \dot{U}} < 0$ 外稳幅条件 非线性反馈网络

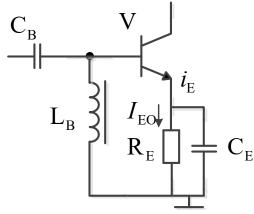
稳定: θ≥90° Soft Self Excitation 软自激

θ<90° Hard Self Excitation 硬自激

自动起振: 电源接通瞬间A类(θ =180°)

$$U_{\rm i} \uparrow U_{\rm o} \uparrow \theta \downarrow$$





发射极自给偏置电路

2. 相位稳定条件

$$U = U_m \sin(\omega t + \varphi_0)$$

$$\omega' = \frac{d\varphi(t)}{dt} = \frac{d(\omega t + \varphi_0)}{dt}$$

$$\varphi_0$$
: 常数 $\omega' = \omega$

$$\phi_0$$
: 常数 $\omega' = \omega$

$$\phi_0$$
: 非常数 $\omega' = \omega + \frac{d\phi_0}{dt} = \omega + \Delta\omega$

$$: \varphi \uparrow \qquad \omega \uparrow \qquad f \uparrow$$

相位超前

频率增大

$$\phi_{\Sigma} = 0 \qquad \phi_{\Sigma} \downarrow \qquad 稳定$$

$$\phi_{\Sigma} \uparrow \longrightarrow \phi_{\Sigma} \uparrow \qquad \pi \& E$$

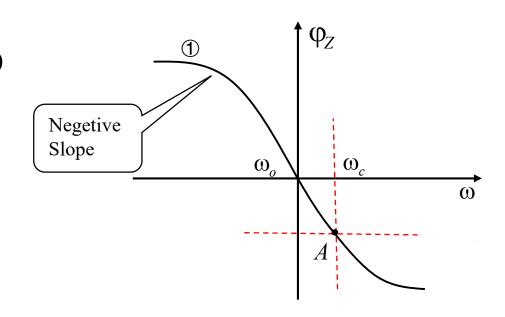
$$\phi_{\Sigma} \downarrow \qquad \phi_{\Sigma} \uparrow \qquad \pi \& E$$

$$\phi_{\Sigma} \downarrow \longrightarrow \phi_{\Sigma} \downarrow \qquad \phi_{\Sigma} \downarrow \qquad \pi \& E$$

相位稳定条件:
$$\frac{\partial \varphi_{\Sigma}}{\partial \omega}\Big|_{p} = \left(\frac{\partial \varphi_{Y}}{\partial \omega}\Big|_{p} + \frac{\partial \varphi_{F}}{\partial \omega}\Big|_{p} + \frac{\partial \varphi_{Z}}{\partial \omega}\Big|_{p}\right) < 0$$

窄带可认为:
$$\frac{\partial \varphi_F}{\partial \omega} \approx 0$$
 $\frac{\partial \varphi_Y}{\partial \omega} \approx 0$

相位稳定条件: $\frac{\partial \varphi_Z}{\partial \omega}\Big|_p < 0$



◆ 5.1.3 起振条件

起始 U_i 小,工作于A类,自起振

: 起振用小信号微变等效电路分析法

起始状态

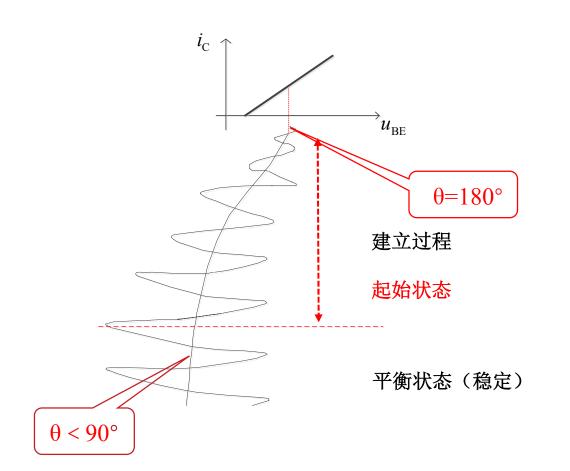
 $\overrightarrow{AF} > 1$

工作于 A 类, θ≥90° 小信号、线性 微变等效电路分析法

稳定状态

 $\overrightarrow{AF} = 1$

工作于 C 类, θ<90° 大信号、非线性 不能用微变等效电路分析法 Y参数等效电路分析法



 $\overrightarrow{AF} > 1$

工作于 A 类, θ≥90° 小信号、线性 微变等效电路分析法

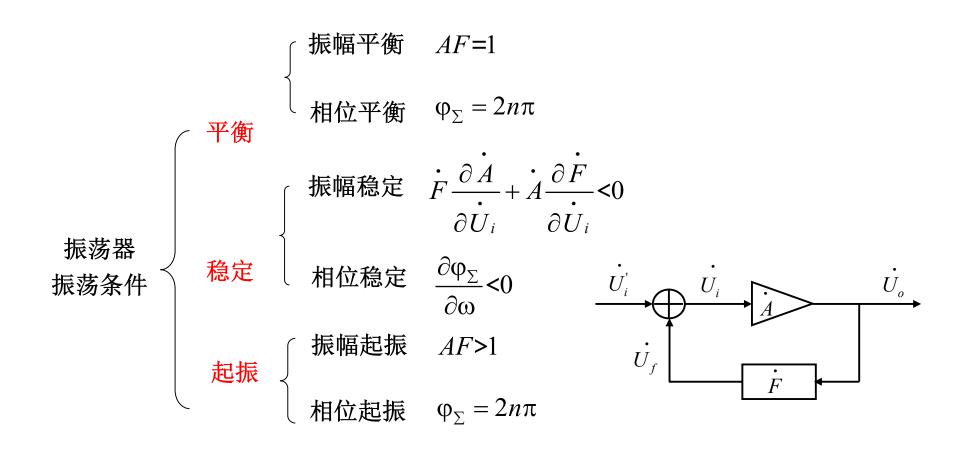
 AF=1

 工作于 C 类, θ<90°</td>

 大信号、非线性

 不能用微变等效电路分析法

 Y参数等效电路分析法



放大器: BJT、FET、差分放大器、运算放大器等

振荡器 组成 反馈网络: RC 移相、电容分压、电感分压、变压器耦合或 电阻分压网络等

选频网络: LC、RC、晶体滤波器等

谢谢!